XC9103/XC9104/XC9105 Series

Ceramic Capacitor Compatible, Step-up DC/DC Controllers

GreenOperation Compatible

GENERAL DESCRIPTION

The XC9103/XC9104/XC9105 series are PWM, PWM/PFM auto switching /manual switching controlled universal step-up DC/DC converter controllers.

Output will be stable no matter which load capacitors are used but should a low ESR capacitor be used, R_{SENSE} of about 0.1 will be required and phase compensation will be achieved. This allows the use of ceramic capacitors and enables to obtain lower output ripple and small PCB design. Tantalum and electrolytic capacitors can also be used, in which case, R_{SENSE} becomes unnecessary.

With 0.9V internal voltage reference and by using externally connected two resistors, output voltage can be set freely within a range of 1.5V to 30V. The series is available in 300 kHz and 180 kHz frequencies, the size of the external components can be reduced. 100 kHz and 500 kHz are also available in custom options.

The XC9103 offers PWM operation. The XC9104 offers PWM/PFM automatic switching operation. The PWM operation is shifted to the PFM operation automatically at light load so that it maintains high efficiency over a wide range of load currents. The XC9105 offers both PWM and PWM/PFM auto switching operations and it can be selected by external signal.

A current limiter circuit is built-in to the IC (except with the 500 kHz version) and monitors the ripple voltage on the FB pin. Operation is shut down when the ripple voltage is more than 250mV. The operations of the IC can be returned to normal with a toggle of the CE pin or by turning the power supply back on.

APPLICATIONS

PDAs

Cellular phones

Palmtop computers

Portable audio systems

Various multi-function power supplies

FEATURES

|	
Input Voltage Range	: 0.9V ~ 10V
Supply Voltage Range	: 1.8V ~ 10V
Output Voltage Range	: 1.5V ~ 30V
	Set freely with the reference voltage 0.9V($\pm$ 2.0%) and two resistors
Oscillation Frequency	: 100, 180, 300, 500kHz ($\pm$ 15%) 180, 300kHz only for XC9103/04/05B type (with current limiter)
Output Current	: more than 400mA (V_{IN}=1.8V, V_{OUT}=3.3V)
Controls	: PWM (XC9103) PWM/PFM auto-switching (XC9104) PWM/PFM manual switching (XC9105)
High Efficiency	: 85% (TYP.)
Stand-by Current	: I_{STB}=1.0μ A (MAX.)
Load Capacitors	: Low ESR capacitors compatible
Current	: Operates when
Limiter Function	ripple voltage=250mV Also available without current limiter (100kHz and 500kHz types are available only without current limiter)
Packages	: SOT-25, USP-6B
Environmentally Friendly	: EU RoHS Compliant, Pb Free

TYPICAL APPLICATION CIRCUIT

TYPICAL PERFORMANCE CHARACTERISTICS

XC9105D093MR
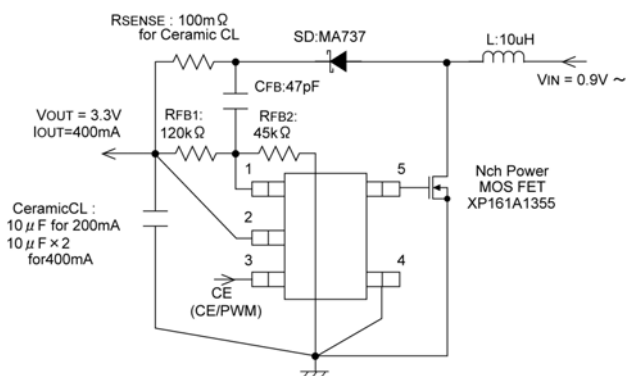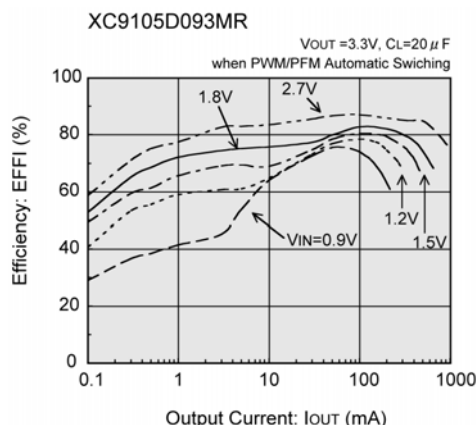
PIN CONFIGURATION
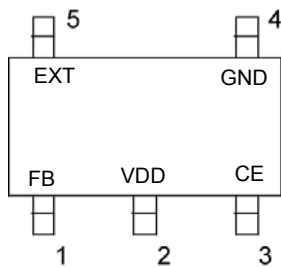
SOT-25
(TOP VIEW)
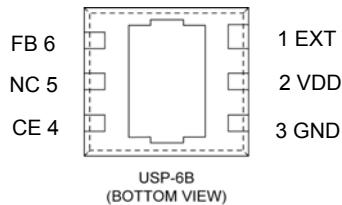
USP-6B
(BOTTOM VIEW)

The dissipation pad for the USP-6B package should be solder-plated in recommended mount pattern and metal masking so as to enhance mounting strength and heat release.
If the pad needs to be connected to other pins, it should be connected to the V_{DD} (No.2) pin.

PIN ASSIGNMENT

PIN NUMBER		PIN NAME	FUNCTION
SOT-25	USP-6B		
1	6	FB	Output Resistor Connection
2	2	V_{DD}	Supply Voltage
3	4	CE	Chip Enable
		CE (/PWM)	Serves as both PWM/PFM switching pin and CE pin for XC9105
4	3	GND	Ground
5	1	EXT	External Transistor Connection
-	5	NC	No Connection

PRODUCT CLASSIFICATION

Ordering Information

XC9103   - ^(*1): PWM Control
XC9104   - ^(*1): PWM/PFM Automatic Switching Control
XC9105   - ^(*1): PWM/PFM Manual Switching Control

DESIGNATOR	DESCRIPTION	SYMBOL	DESCRIPTION
①	Type of DC/DC Controller	B	With current limiter (180kHz, 300kHz only)
		D	Without current limiter
②③	Output Voltage	09	FB voltage (e.g. FB Voltage=0.9V→②=0, ③=9)
④	Oscillation Frequency	3	300kHz
		1	100kHz
		2	180kHz
		5	500kHz
⑤⑥-⑦	Packages Taping Type ^(*2)	MR	SOT-25
		MR-G	SOT-25 (Halogen & Antimony free)
		DR	USP-6B

(*1) The "-G" suffix indicates that the products are Halogen and Antimony free as well as being fully RoHS compliant.
(*2) The device orientation is fixed in its embossed tape pocket. For reverse orientation, please contact your local Torex sales office or representative. (Standard orientation: R-   , Reverse orientation: L-   )

BLOCK DIAGRAM
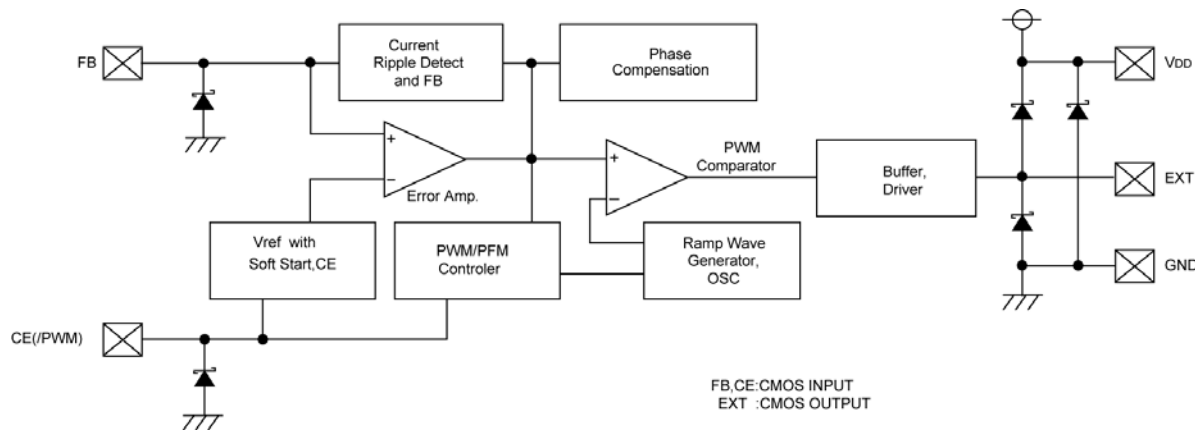
ABSOLUTE MAXIMUM RATINGS

Ta=25

PARAMETER		SYMBOL	RATINGS	UNITS
V_{DD} pin Voltage		V_{DD}	-0.3 ~ 12.0	V
FB pin Voltage		FB	-0.3 ~ 12.0	V
CE pin Voltage		V_{CE}	-0.3 ~ 12.0	V
EXT pin Voltage		V_{EXT}	-0.3 ~ V_{DD} + 0.3	V
EXT pin Current		$I_{EXT/}$	±100	mA
Power Dissipation	SOT-25	Pd	150	mW
	USP-6B		100	
Operating Temperature Range		Topr	-40 ~ +85	
Storage Temperature Range		Tstg	-40 ~ +125	

ELECTRICAL CHARACTERISTICS

XC9103B092MR, XC9104B092MR, XC9105B092MR
XC9103D092MR, XC9104D092MR, XC9105D092MR

(f_{OSC}=180kHz)   Ta=25

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Output Voltage	V_{OUT}		3.234	3.300	3.366	V	
Output Voltage Range	V_{OUTSET}	V_{IN}=V_{OUTSET}x0.6, V_{DD}=3.3V I_{OUT}=10mA, Using 2SD1628	1.5	-	30.0	V	
FB Control Voltage	V_{FB}		0.882	0.900	0.918	V	
Supply Voltage Range (*1)	V_{DD}		1.8	-	10.0	V	
Operation Start Voltage	V_{ST1}	Recommended circuit using 2SD1628, I_{OUT}=1.0mA	-	-	0.9	V	
Oscillation Start Voltage (*1)	V_{ST2}	No external components, CE connected to V_{DD}, Voltage applied, FB=0V	-	-	0.8	V	
Operation Hold Voltage	V_{HLD}	Recommended circuit using 2SD1628, I_{OUT}=1.0mA	-	-	0.7	V	
Supply Current 1	I_{DD1}	Same as V_{ST2}, V_{DD}=3.3V	-	45	64	μA	
Supply Current 2	I_{DD2}	Same as I_{DD1}, FB=1.2V	-	17	24	μA	
Stand-by Current	I_{STB}	Same as I_{DD1}, CE=0V	-	-	1.0	μA	
Oscillation Frequency	f_{OSC}	Same as I_{DD1}	153	180	207	kHz	
Maximum Duty Cycle	MAXDTY	Same as I_{DD1}	75	81	87	%	
PFM Duty Rate	PFMDTY	No load (XC9104B/D, 9105B/D)	20	28	36	%	
Overcurrent Sense Voltage (*3)	V_{LMT}	Step input to FB (Pulse width: 2.0μS or more), EXT=Low level voltage (XC9103B, 9104B, 9105B)	170	250	330	mV	
Efficiency	EFFI	Recommended circuit using XP161A1355	-	85	-	%	
Soft-Start Time	t_{SS}		5.0	10.0	20.0	ms	
CE "High" Voltage (*2)	V_{CEH}	Same as I_{DD1}	0.65	-	-	V	
CE "Low" Voltage (*2)	V_{CEL}	Same as I_{DD1}	-	-	0.20	V	
PWM "High" Voltage (*2)	V_{PWMH}	I_{OUT}=1.0mA (XC9105B/D)	V_{DD}-0.2	-		V	
PWM "Low" Voltage (*2)	V_{PWML}	I_{OUT}=1.0mA (XC9105B/D)	-	-	V_{DD}-1.0	V	
EXT "High" On Resistance	R_{EXTH}	Same as I_{DD1}, V_{EXT}=V_{OUT}-0.4V	-	24	36		
EXT "Low" On Resistance	R_{EXTL}	Same as I_{DD1}, V_{EXT}=0.4V	-	16	24		
CE "High" Current	I_{CEH}	Same as I_{DD2}, CE=V_{DD}	-	-	0.1	μA	
CE "Low" Current	I_{CEL}	Same as I_{DD2}, CE=0V	-	-	- 0.1	μA	
FB "High" Current	I_{FBH}	Same as I_{DD2}, CE=V_{DD}	-	-	0.1	μA	
FB "Low" Current	I_{FBL}	Same as I_{DD2}, CE=0V	-	-	- 0.1	μA	

Test Conditions: Unless otherwise stated, CL: ceramic, recommended MOSFET should be connected.
        V_{OUT}=3.3V, V_{IN}=2.0V, I_{OUT}=170mA

NOTE:
*1 Although the IC starts step-up operations from a V_{DD} of 0.8V, the output voltage and oscillation frequency are stabilized at $V_{DD} \geq$1.8V. Therefore, a V_{DD} of more than 1.8V is recommended when V_{DD} is supplied from V_{IN} or other power sources.
*2 With the XC9105 series, the CE pin also serves as a PWM/PFM switching pin. In operation, PWM control is selected when the voltage at the CE pin is more than V_{DD} -0.2V.   On the other hand, PWM/PFM automatic switching control at a duty = 25% is selected when the voltage at the CE pin is less than V_{DD} -1.0V and more than V_{CEH}.
*3 The overcurrent limit circuit of this IC is designed to monitor the ripple voltage so please select your external components carefully to prevent V_{LMT} being reached under low temperature conditions as well as normal operating conditions.   Following current limiter circuit operations, which in turn causes the IC's operations to stop, the operations of the IC can be returned to normal with a toggle of the CE pin or by turning the power supply back on.

ELECTRICAL CHARACTERISTICS (Continued)

XC9103B093MR, XC9104B093MR, XC9105B093MR
XC9103D093MR, XC9104D093MR, XC9105D093MR

(f_{OSC}=300 kHz)   Ta=25

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Output Voltage	V_{OUT}		3.234	3.300	3.366	V	
Output Voltage Range	V_{OUTSET}	$V_{IN}=V_{OUTSET}$x0.6, V_{DD}=3.3V I_{OUT}=10mA, Using 2SD1628	1.5	-	30.0	V	
FB Control Voltage	V_{FB}		0.882	0.900	0.918	V	
Supply Voltage Range (*1)	V_{DD}		1.8	-	10.0	V	
Operation Start Voltage	V_{ST1}	Recommended circuit using 2SD1628, I_{OUT}=1.0mA	-	-	0.9	V	
Oscillation Start Voltage (*1)	V_{ST2}	No external components, CE connected to V_{DD}, Voltage applied, FB=0V	-	-	0.8	V	
Operation Hold Voltage	V_{HLD}	Recommended circuit using 2SD1628, I_{OUT}=1.0mA	-	-	0.7	V	
Supply Current 1	I_{DD1}	Same as V_{ST2}, V_{DD}=3.3V	-	62	88	μ A	
Supply Current 2	I_{DD2}	Same as I_{DD1}, FB=1.2V	-	16	22	μ A	
Stand-by Current	I_{STB}	Same as I_{DD1}, CE=0V	-	-	1.0	μ A	
Oscillation Frequency	f_{OSC}	Same as I_{DD1}	255	300	345	kHz	
Maximum Duty Cycle	MAXDTY	Same as I_{DD1}	75	81	87	%	
PFM Duty Rate	PFMDTY	No load (XC9104B/D, 9105B/D)	24	32	40	%	
Overcurrent Sense Voltage (*3)	V_{LMT}	Step input to FB (Pulse width: 2.0μ S or more), EXT=Low level voltage (XC9103B, 9104B, 9105B)	220	300	380	mV	
Efficiency	EFFI	Recommended circuit using XP161A1355	-	85	-	%	
Soft-Start Time	t_{SS}		5.0	10.0	20.0	ms	
CE "High" Voltage (*2)	V_{CEH}	Same as I_{DD1}	0.65	-	-	V	
CE "Low" Voltage (*2)	V_{CEL}	Same as I_{DD1}	-	-	0.20	V	
PWM "High" Voltage (*2)	V_{PWMH}	I_{OUT}=1.0mA (XC9105B/D)	V_{DD}-0.2	-	-	V	
PWM "Low" Voltage (*2)	V_{PWML}	I_{OUT}=1.0mA (XC9105B/D)	-	-	V_{DD}-1.0	V	
EXT "High" On Resistance	R_{EXTH}	Same as I_{DD1}, $V_{EXT}=V_{OUT}$-0.4V	-	24	36		
EXT "Low" On Resistance	R_{EXTL}	Same as I_{DD1}, V_{EXT}=0.4V	-	16	24		
CE "High Current	I_{CEH}	Same as I_{DD2}, CE=V_{DD}	-	-	0.1	μ A	
CE "Low" Current	I_{CEL}	Same as I_{DD2}, CE=0V	-	-	- 0.1	μ A	
FB "High" Current	I_{FBH}	Same as I_{DD2}, CE=V_{DD}	-	-	0.1	μ A	
FB "Low" Current	I_{FBL}	Same as I_{DD2}, CE=0V	-	-	- 0.1	μ A	

Test Conditions: Unless otherwise stated, CL: ceramic, recommended MOSFET should be connected.
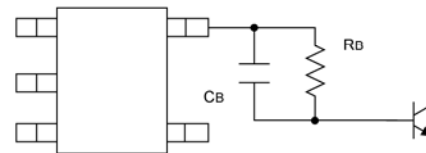
NOTE:
*1 Although the IC starts step-up operations from a V_{DD} of 0.8V, the output voltage and oscillation frequency are stabilized at $V_{DD} \geq$1.8V. Therefore, a V_{DD} of more than 1.8V is recommended when V_{DD} is supplied from V_{IN} or other power sources.
*2 With the XC9105 series, the CE pin also serves as a PWM/PFM switching pin. In operation, PWM control is selected when the voltage at the CE pin is more than V_{DD} -0.2V.   On the other hand, PWM/PFM automatic switching control at a duty = 25% is selected when the voltage at the CE pin is less than V_{DD} -1.0V and more than V_{CEH}.
*3 The overcurrent limit circuit of this IC is designed to monitor the ripple voltage so please select your external components carefully to prevent V_{LMT} being reached under low temperature conditions as well as normal operating conditions. Following current limiter circuit operations, which in turn causes the IC's operations to stop, the operations of the IC can be returned to normal with a toggle of the CE pin or by turning the power supply back on.

TYPICAL APPLICATION CIRCUIT
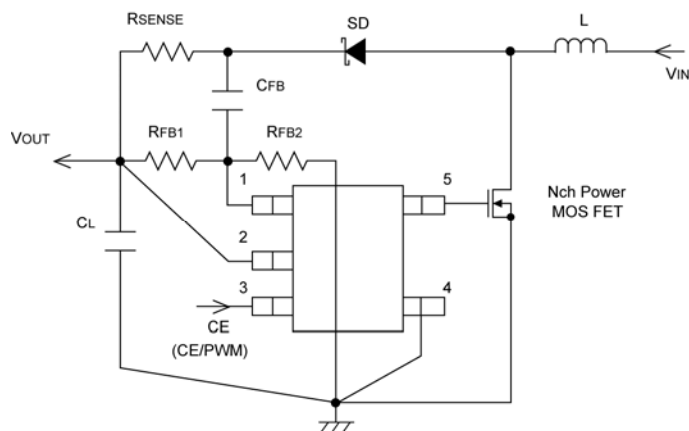
When obtaining V_{DD} from a source other than V_{OUT}, please insert a capacitor C_{DD} between the V_{DD} pin and the GND pin in order to provide stable operations.
Please place C_L and C_{DD} as close as to the V_{OUT} and V_{DD} pins respectively and also close to the GND pin. Strengthen the wiring sufficiently. RSENSE should be removed and shorted when the C_L capacitor except for ceramic or low ESR capacitor is used.

Insert R_B and C_B when using a bipolar NPN Transistor.

NOTES ON USE

<XC9105 CE/PWM PIN>
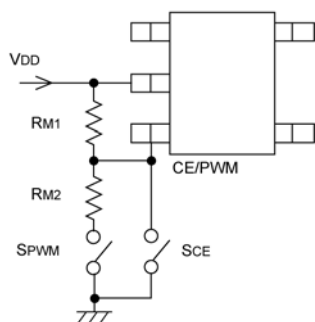
S_{CE}	S_{PWM}	CONDITIONS
ON		Chip Disable
OFF	ON	Duty=25%, PWM/PFM automatic switching
OFF	OFF	PWM

By using external signals, the control of the XC9105 series can be alternated between PWM control and PWM/PFM automatic switching control.   By inputting a voltage of more than V_{DD} -0.2V to the CE/PWM pin, PWM control can be selected.   On the other hand, PWM/PFM automatic switching control can be selected by inputting a voltage of less than V_{DD} -1.0V.
With the XC9105, by connecting resistors of the same value (R_{M1}, R_{M2}) as shown in the diagram to the left, it is possible to obtain chip disable with S_{CE} ON and, S_{PWM} ON or OFF, PWM/PFM auto switching at Duty=25% with S_{CE} OFF & S_{PWM} ON, & PFM control with both switches OFF.

Note:
When operating at V_{DD} -1.8V and below (stepping-up from VIN=0.9V), it is necessary to pull-up to V_{DD} in order to allow the CE/PWM pin reach the V_{CEH} voltage level.   Please make sure that the IC is in PWM control (S_{PWM}=OFF) when operations start. If S_{PWM} is ON, there are times when chip enable might not operate.

 * Please select your external components carefully.

OPERATIONAL EXPLANATION

The XC9103/04/05 series are step-up DC/DC converter controller ICs with built-in high speed, low ON resistance drivers.

<Error Amp.>
Error amplifier is designed to monitor the output voltage, comparing the feedback voltage (FB) with the reference voltage Vref.   In response to feedback of a voltage lower than the reference voltage Vref, the output voltage of the error amp. decreases.

<OSC Generator>
This circuit generates the internal reference clock.

<Ramp Wave Generator>
The ramp wave generator generates a saw-tooth waveform based on outputs from the OSC Generator.

<PWM Comparator>
The PWM comparator compares outputs from the error amp. and saw-tooth waveform.   When the voltage from the Error Amp's output is low, the external switch will be set to ON.

<PWM/PFM Controller>
This circuit generates PFM pulses.
The PWM/PFM automatic switching mode switches between PWM and PFM automatically depending on the load.   The PWM/PFM automatic switching mode is selected when the voltage of the CE pin is less than V_{DD} - 1.0V, and the control switches between PWM and PFM automatically depending on the load.   PWM/PFM control turns into PFM control when threshold voltage becomes lower than voltage of error amps.   PWM control mode is selected when the voltage of the CE pin is more than V_{DD} - 0.2V. Noise is easily reduced with PWM control since the switching frequency is fixed. The series is suitable for noise sensitive portable audio equipment as PWM control can suppress noise during operation and PWM/PFM switching control can reduce consumption current during light load in stand-by.

<Vref 1 with Soft Start>
The reference voltage, Vref (FB pin voltage)=0.9V, is adjusted and fixed by laser trimming (for output voltage settings, please refer to the notes on next page). To protect against inrush current, when the power is switched on, and also to protect against voltage overshoot, soft-start time is set internally to 10ms.   It should be noted, however, that this circuit does not protect the load capacitor (C_L) from inrush current. With the Vref voltage limited and depending upon the input to the error amps, the operation maintains a balance between the two inputs of the error amps and controls the EXT pin's ON time so that it doesn't increase more than is necessary.

<Enable Function>
This function controls the operation and shutdown of the IC.   When the voltage of the CE pin is 0.2V or less, the mode will be disable, the channel's operations will stop and the EXT1 pin will be kept at a low level (the external N-type MOSFET will be OFF).   When the IC is in a state of disable, current consumption will be no more than 1.0μ A.
When the CE pin's voltage is 0.65V or more, the mode will be enabled and operations will recommence.

OPERATIONAL EXPLANATION (Continued)

< Output Voltage Setting>
Output voltage can be set by adding external split resistors.   Output voltage is determined by the following equation, based on the values of R_{FB1} and R_{FB2}.   The sum of R_{FB1} and R_{FB2} should normally be 2 M or less.

$$V_{OUT} = 0.9 \times (R_{FB1} + R_{FB2}) / R_{FB2}$$

The value of C_{FB1}, speed-up capacitor for phase compensation, should result in fzfb = $1/(2 \times C_{FB} \times R_{FB1})$ equal to 5 to 30kHz.   Adjustments are required depending on the application, value of inductance (L), and value of load capacity (CL).

fzfb = 30kHz (L=10μH)	[Example of Equation]
fzfb = 20kHz (L=22μH)	R_{FB1} : 120k R_{FB2} : 45k
fzfb = 10kHz (L=47μH)	C_{FB} : 47pF (fzfb = 30kHz, L = 10μH)
	68pF (fzfb = 20kHz, L = 22μH)
	130pF (fzfb = 10kHz, L = 47μH)

< The use of ceramic capacitor CL >
The circuit of the XC9103/04/05 series is organized by a specialized circuit, which reenacts negative feedback of both voltage and current.   Also by insertion of approximately 100m of a low and inexpensive sense resistor as current sense, a high degree of stability is possible even using a ceramic capacitor, a condition which used to be difficult to achieve. Compared to a tantalum condenser, because the series can be operated in a very small capacity, it is suited to use of the ceramic capacitor, which is cheap and small.

< External Components >

Tr   :*When a MOSFET is used:
        XP161A1355PR (N-ch Power MOSFET, TOREX)
        Note*: As the breakdown voltage of XP161A1355 is 8V, take care
               with the power supply voltage.
               With output voltages over 6V, use the XP161A1265 with a
               breakdown voltage of 12V.
        VST1: XP161A1355PR =1.2V (MAX.)
               XP161A1265PR = 1.5V (MAX.)
SD   :MA2Q737 (Schottky type, MATSUSHITA)
L, C_L   :**When Using Ceramic Type**
L   :22μH (CDRH5D28, SUMIDA, f_{OSC} = 100, 180kHz)
        10μH (CDRH5D18, SUMIDA, f_{OSC} = 300, 500kHz)
C_L :10V 10μF (Ceramic Type, LMK325BJ106ML, TAIYO YUDEN)
           Use the formula below when step-up ratio and output current is
           large.
           $CL = (CL$ standard value$) \times (I_{OUT}(mA) / 300mA \times V_{OUT} / V_{IN})$
R_{SENSE} :100m   (f_{OSC} = 180, 300, 500kHz)
        50m   (f_{OSC} = 100kHz)

C_L   :**Tantalum Type**
L   :22μH (CDRH5D28, SUMIDA, f_{OSC} = 300kHz)
        47μH (CDRH5D28, SUMIDA, f_{OSC} = 100, 180kHz)
           Except when $I_{OUT}(mA) / 100mA \times V_{OUT} / V_{IN} > 2$ 22μH
        10μH (CDRH5D18, SUMIDA, f_{OSC} = 500kHz)
C_L :16V, 47μF (Tantalum Type 16MCE476MD2, NICHICHEMI)
           Use the formula below when step-up ratio and output current is
           large.
           $CL = (CL$ standard value$) \times (I_{OUT}(mA) / 300mA \times V_{OUT} / V_{IN})$
R_{SENSE} :Not required, but short out the wire.

C_L   :**AL Electrolytic Type**
L   :22μH (CDRH5D28 SUMIDA, f_{OSC} = 300kHz)
        47μH (CDRH5D28 SUMIDA, f_{OSC} = 100, 180kHz)
           Except when $I_{OUT}(mA) / 100mA \times V_{OUT} / V_{IN} > 2$ 22μH
C_L :16V, 100μF (AL Electrolytic Type) + 10V, 2.2μF (Ceramic Type)
           Strengthen appropriately when step-up ratio and output current is
           large.
R_{SENSE} :Not required, but short out the wire.
CFB   :Set up so that fzfb = 100kHz.

*When a NPN Tr. Is used:

2SD1628 (SANYO)

R_B : 500   (Adjust with Tr's HSE or load)

C_B : 2200pF (Ceramic type set so that R_B and
      pole is less than 70% of f_{OSC})
      $C_B \leq 1 / (2 \times R_B \times f_{OSC} \times 0.7)$

TEST CIRCUITS

Circuit

Circuit
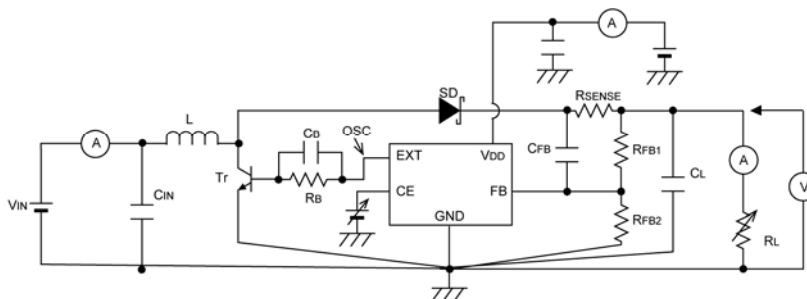
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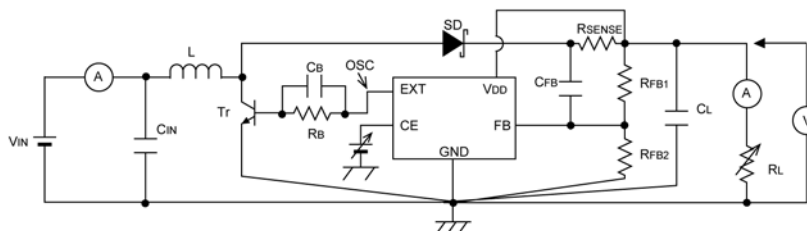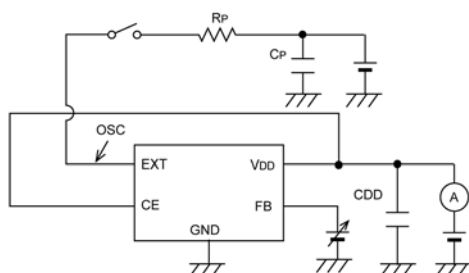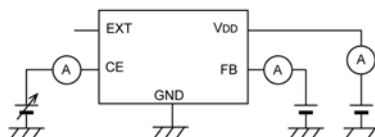
Circuit

Circuit
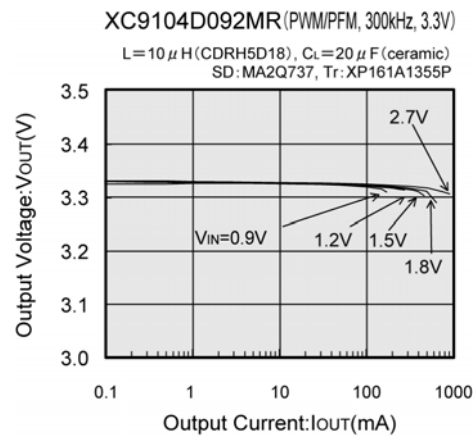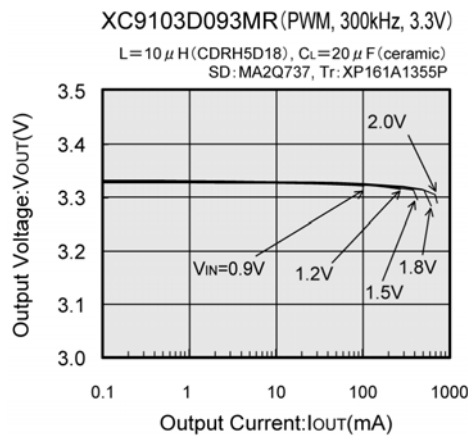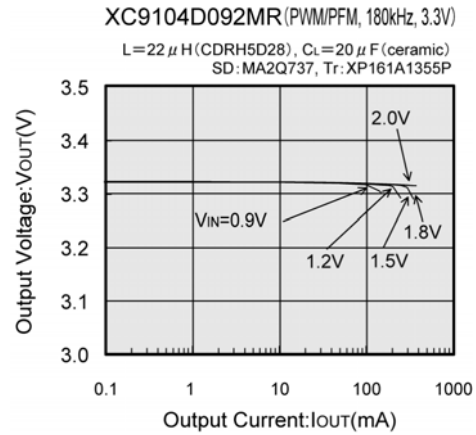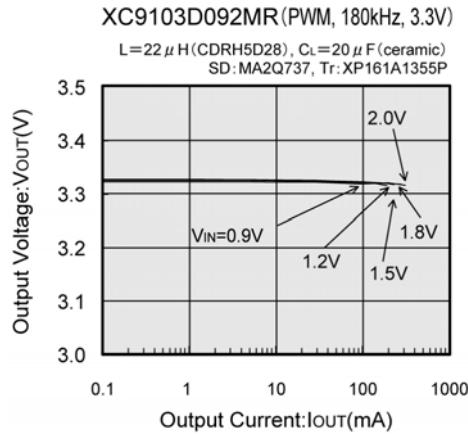

Circuit
Pulse voltage is applied at the FB pin
using the test circuit

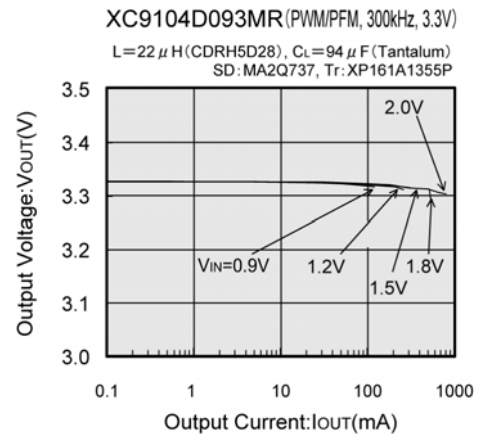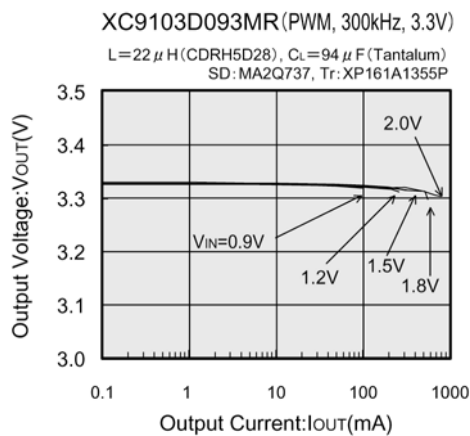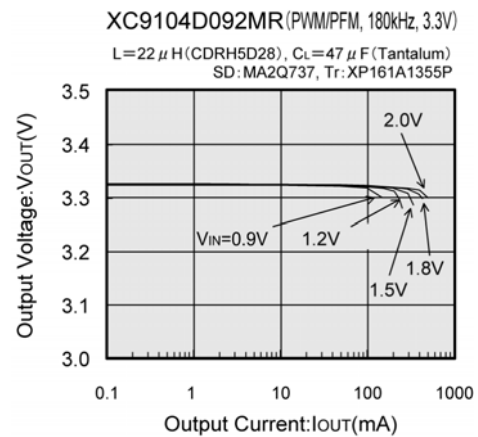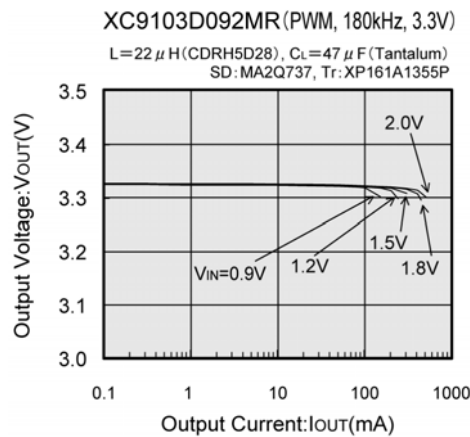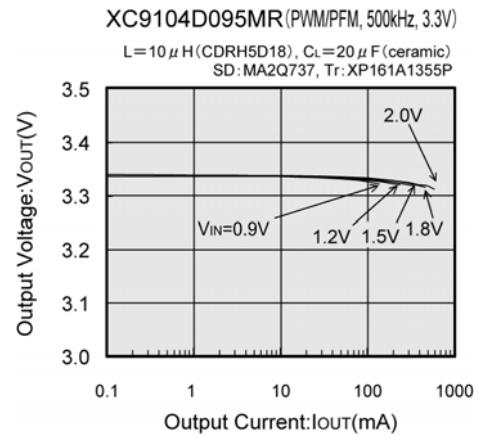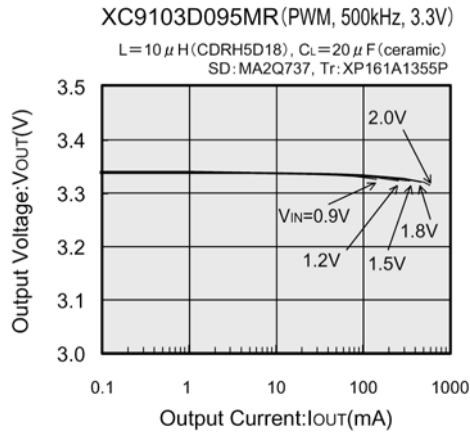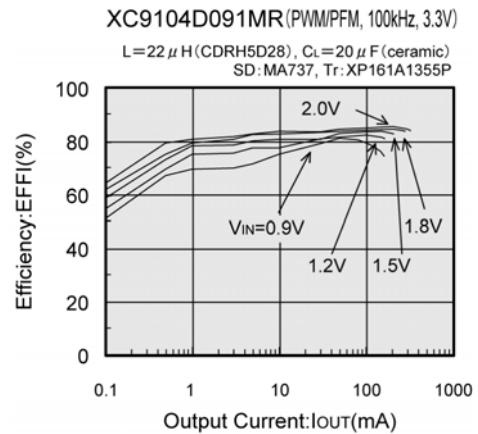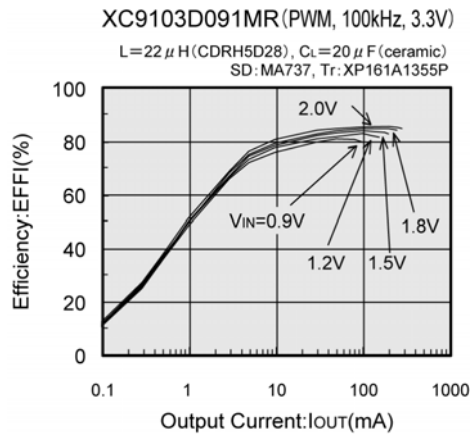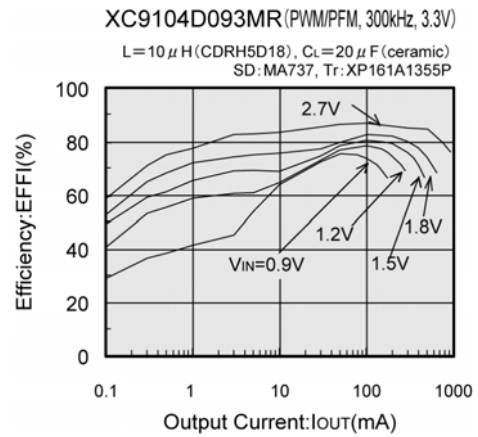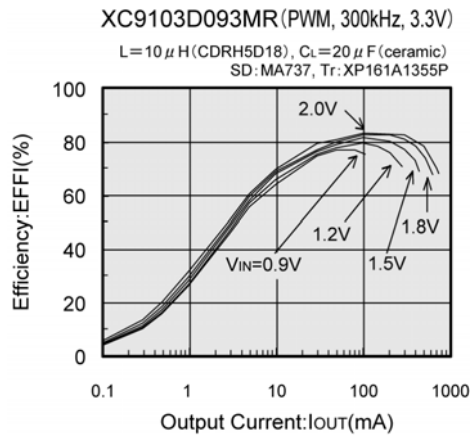## TYPICAL PERFORMANCE CHARACTERISTICS
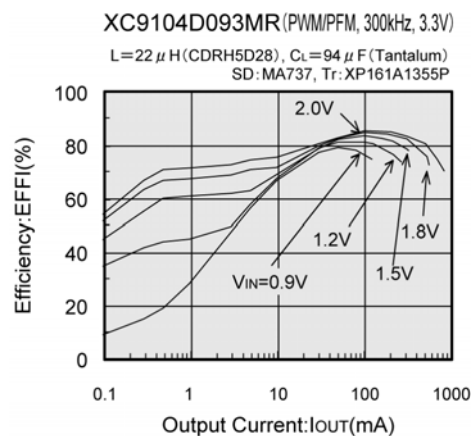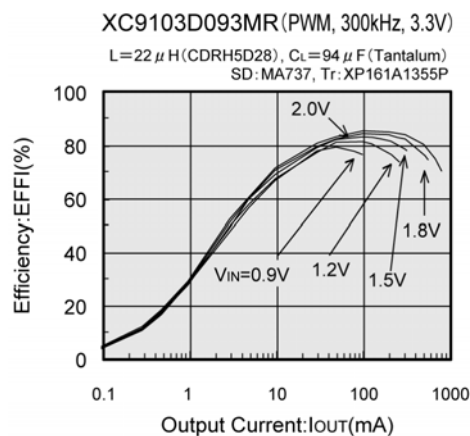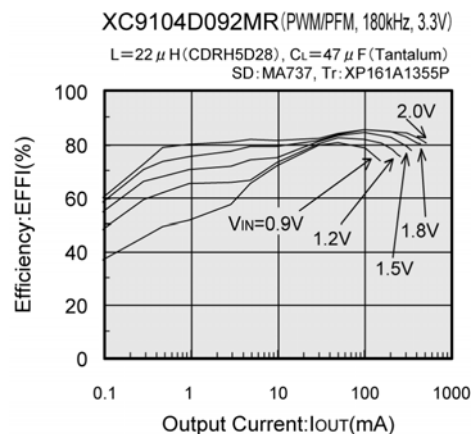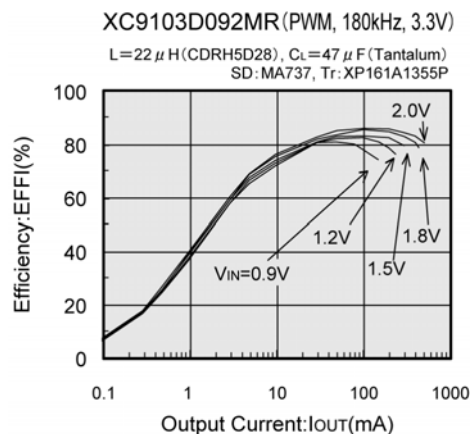
(1) Output Voltage vs. Output Current

XC9103D092MR (PWM, 180kHz, 3.3V)

L=22μH (CDRH5D28), C$_L$=20μF (ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)

L=22μH (CDRH5D28), C$_L$=20μF (ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9103D093MR (PWM, 300kHz, 3.3V)

L=10μH (CDRH5D18), C$_L$=20μF (ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9104D092MR (PWM/PFM, 300kHz, 3.3V)

L=10μH (CDRH5D18), C$_L$=20μF (ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9103D091MR (PWM, 100kHz, 3.3V)

L=22μH (CDRH5D28), C$_L$=20μF (ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9104D091MR (PWM/PFM, 100kHz, 3.3V)

L=22μH (CDRH5D28), C$_L$=20μF (ceramic)
SD:MA2Q737, Tr:XP161A1355P

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(1) Output Voltage vs. Output Current (Continued)

XC9103D095MR(PWM, 500kHz, 3.3V)
$L=10\mu H$(CDRH5D18), $C_L=20\mu F$(ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9104D095MR(PWM/PFM, 500kHz, 3.3V)
$L=10\mu H$(CDRH5D18), $C_L=20\mu F$(ceramic)
SD:MA2Q737, Tr:XP161A1355P

XC9103D092MR(PWM, 180kHz, 3.3V)
$L=22\mu H$(CDRH5D28), $C_L=47\mu F$(Tantalum)
SD:MA2Q737, Tr:XP161A1355P

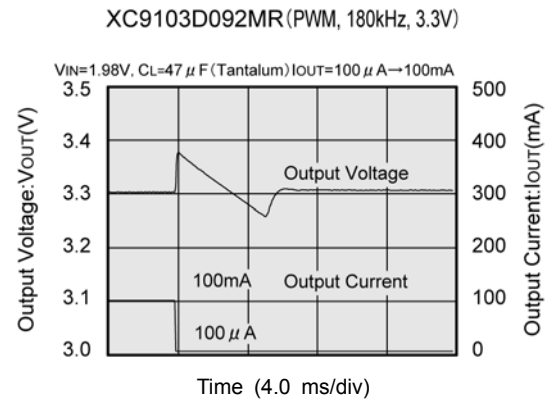### XC9104D092MR(PWM/PFM, 180kHz, 3.3V)
$L=22\mu H$(CDRH5D28), $C_L=47\mu F$(Tantalum)
SD:MA2Q737, Tr:XP161A1355P

XC9103D093MR(PWM, 300kHz, 3.3V)
$L=22\mu H$(CDRH5D28), $C_L=94\mu F$(Tantalum)
SD:MA2Q737, Tr:XP161A1355P

XC9104D093MR(PWM/PFM, 300kHz, 3.3V)
$L=22\mu H$(CDRH5D28), $C_L=94\mu F$(Tantalum)
SD:MA2Q737, Tr:XP161A1355P

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(2) Efficiency vs. Output Current (Continued)

XC9103D092MR (PWM, 180kHz, 3.3V)

L=22μH(CDRH5D28), CL=20μF(ceramic)
SD：MA737, Tr：XP161A1355P

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)

L=22μH(CDRH5D28), CL=20μF(ceramic)
SD：MA737, Tr：XP161A1355P

XC9103D093MR (PWM, 300kHz, 3.3V)

L=10μH(CDRH5D18), CL=20μF(ceramic)
SD：MA737, Tr：XP161A1355P

XC9104D093MR (PWM/PFM, 300kHz, 3.3V)

L=10μH(CDRH5D18), CL=20μF(ceramic)
SD：MA737, Tr：XP161A1355P

XC9103D091MR (PWM, 100kHz, 3.3V)

L=22μH(CDRH5D28), CL=20μF(ceramic)
SD：MA737, Tr：XP161A1355P

XC9104D091MR (PWM/PFM, 100kHz, 3.3V)

L=22μH(CDRH5D28), CL=20μF(ceramic)
SD：MA737, Tr：XP161A1355P

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(2) Efficiency vs. Output Current (Continued)

XC9103D095MR (PWM, 500kHz, 3.3V)
L=10μH (CDRH5D18), CL=20μF (ceramic)
SD:MA737, Tr:XP161A1355P

XC9104D095MR (PWM/PFM, 500kHz, 3.3V)
L=10μH (CDRH5D18), CL=20μF (ceramic)
SD:MA737, Tr:XP161A1355P

XC9103D092MR (PWM, 180kHz, 3.3V)
L=22μH (CDRH5D28), CL=47μF (Tantalum)
SD:MA737, Tr:XP161A1355P

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)
L=22μH (CDRH5D28), CL=47μF (Tantalum)
SD:MA737, Tr:XP161A1355P

XC9103D093MR (PWM, 300kHz, 3.3V)
L=22μH (CDRH5D28), CL=94μF (Tantalum)
SD:MA737, Tr:XP161A1355P

XC9104D093MR (PWM/PFM, 300kHz, 3.3V)
L=22μH (CDRH5D28), CL=94μF (Tantalum)
SD:MA737, Tr:XP161A1355P

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(3) Load Transient Response

XC9103D092MR (PWM, 180kHz, 3.3V)

XC9103D092MR (PWM, 180kHz, 3.3V)

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)

XC9103D093MR (PWM, 300kHz, 3.3V)

XC9103D093MR (PWM, 300kHz, 3.3V)
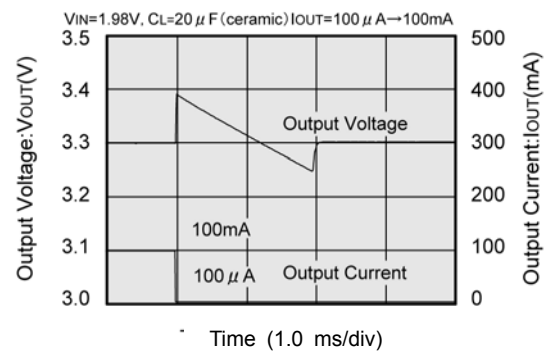
TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(3) Load Transient Response (Continued)

XC9104D093MR (PWM/PFM, 300kHz, 3.3V)

XC9104D093MR (PWM/PFM, 300kHz, 3.3V)

XC9103D092MR (PWM, 180kHz, 3.3V)

XC9103D092MR (PWM, 180kHz, 3.3V)

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)

XC9104D092MR (PWM/PFM, 180kHz, 3.3V)
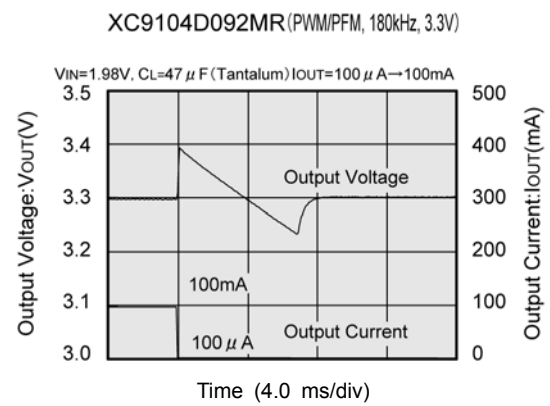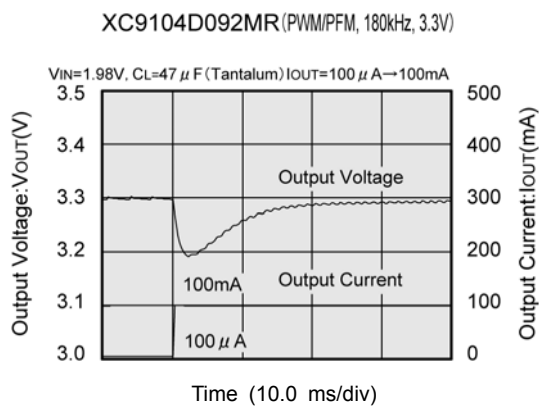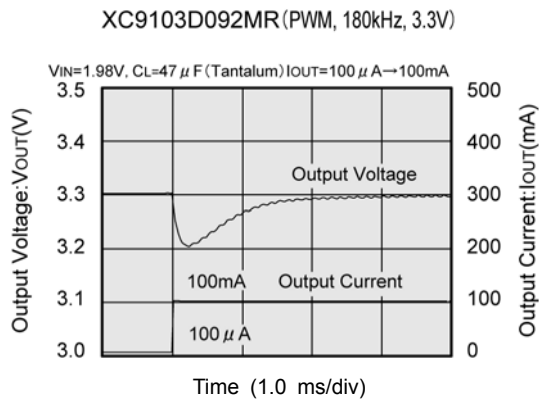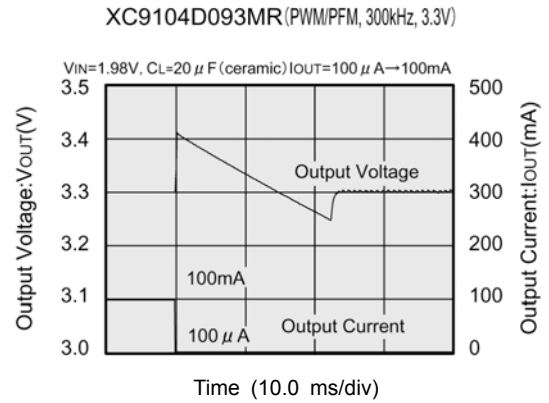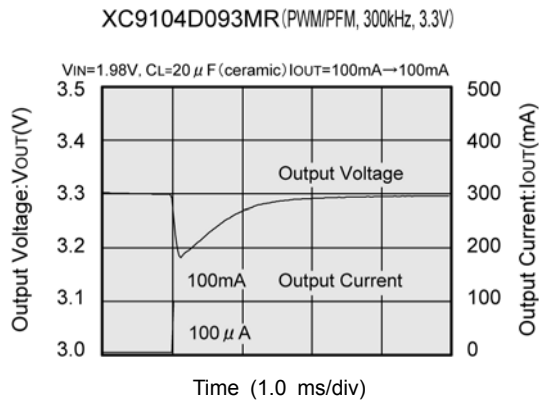
TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(4) Output Voltage vs. Power Supply Voltage

XC9105D092MR（180kHz,3.3V）

L=22μH(CDRH5D28), C_L=20μF(ceramic)
SD：MA2Q737, TXP161A1355P

(5) Supply Current 1 vs. Power Supply Voltage

XC9105D092MR（180kHz）
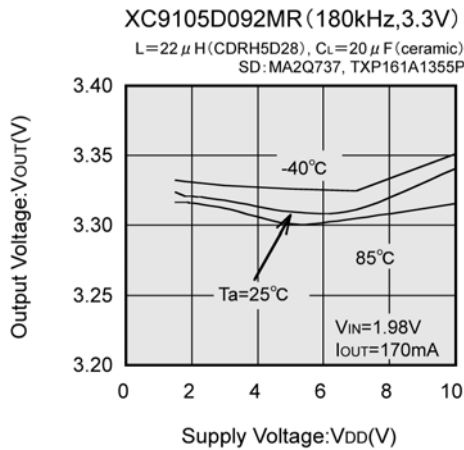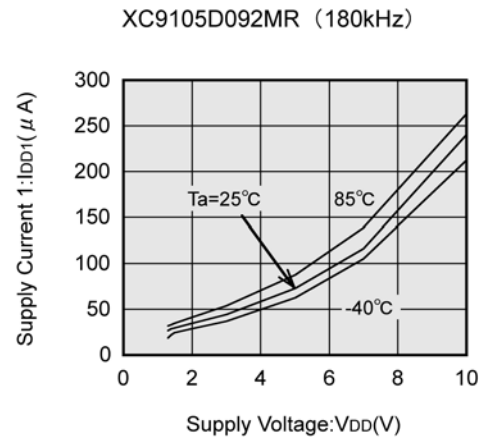
(6) Supply Current 2 vs. Power Supply Voltage

XC9105D092MR(180kHz)
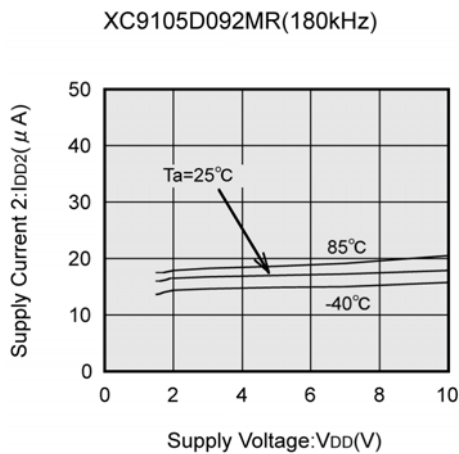
(7) Stand-By Current vs. Power Supply Voltage

XC9105D092MR(180kHz)
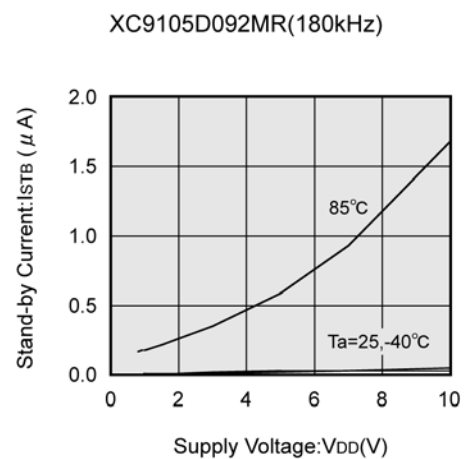
(8) Oscillation Frequency vs. Power Supply Voltage

XC9105D092MR(180kHz)
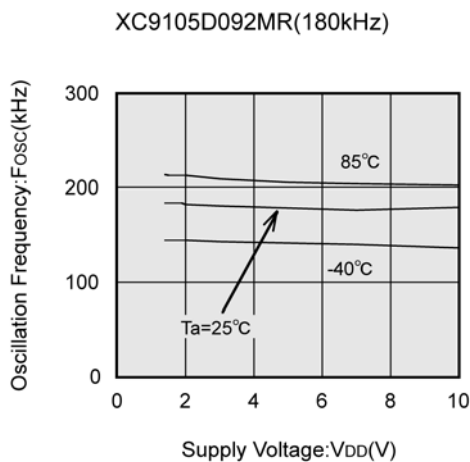
(9) Maximum Duty Ratio vs. Power Supply Voltage

XC9105D092MR(180kHz)
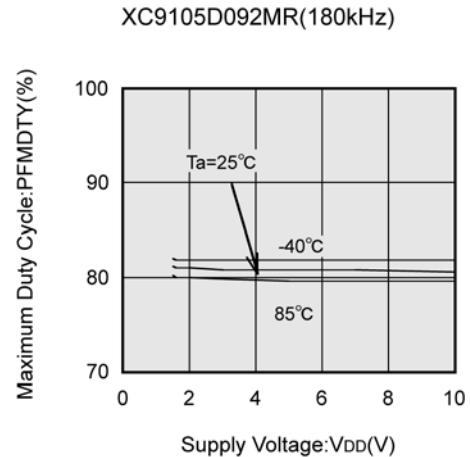
TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(10) PFM Duty Ratio vs. Power Supply Voltage

XC9105D092MR(180kHz)
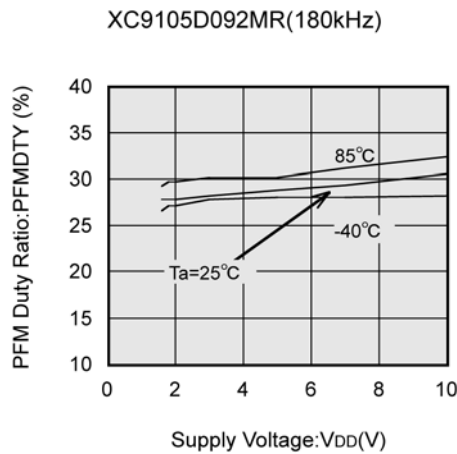
(11) Overcurrent Sense Voltage vs. Power Supply Voltage

XC9105D092MR(180kHz)
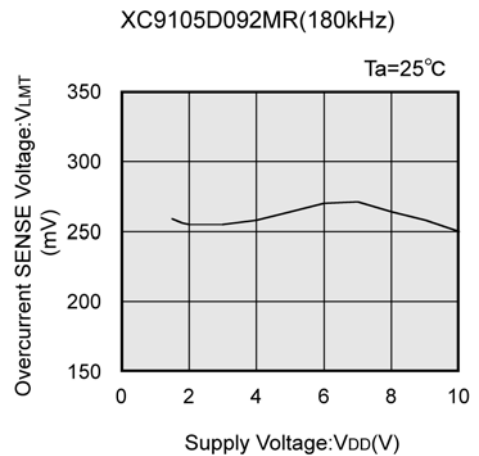
(12) Soft Start Time vs. Power Supply Voltage

XC9105D092MR(180kHz)
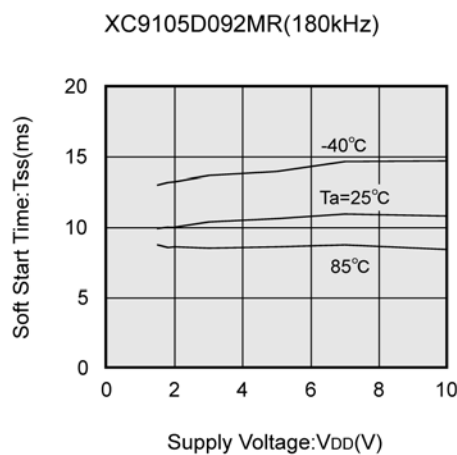
(13) CE "H" "L" Voltage vs. Power Supply Voltage

XC9105D092MR(180kHz)
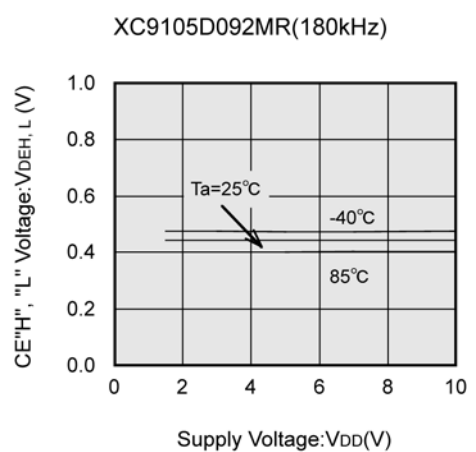
(14) PWM "H" "L" Voltage vs. Power Supply Voltage

XC9105D092MR(180kHz)
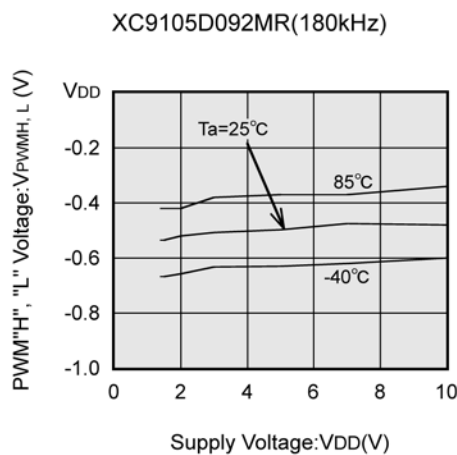
(15) EXT "H" On Resistance vs. Power Supply Voltage

XC9105D092MR(180kHz)
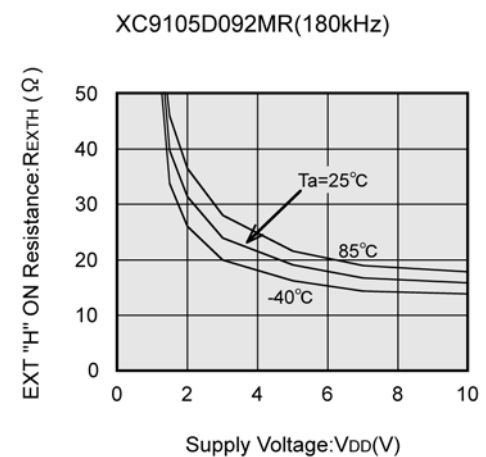
TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(16) EXT "L" On Resistance vs. Power Supply Voltage
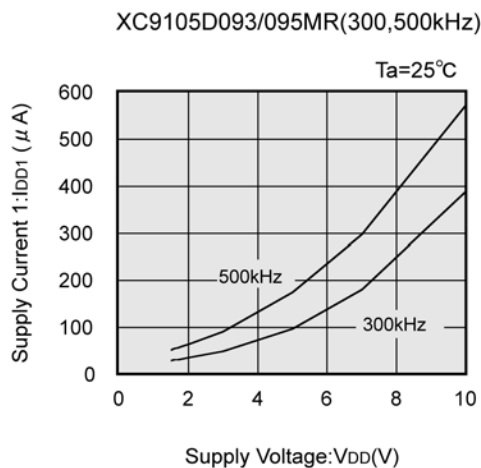
XC9105D092MR(180kHz)

(17) Operation Start Voltage vs. Ambient Temperature

XC9105D092MR (180kHz,3.3V)
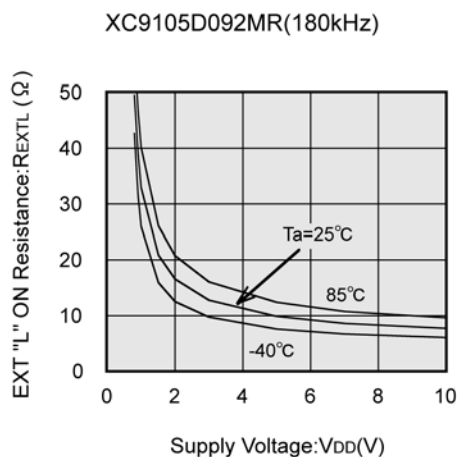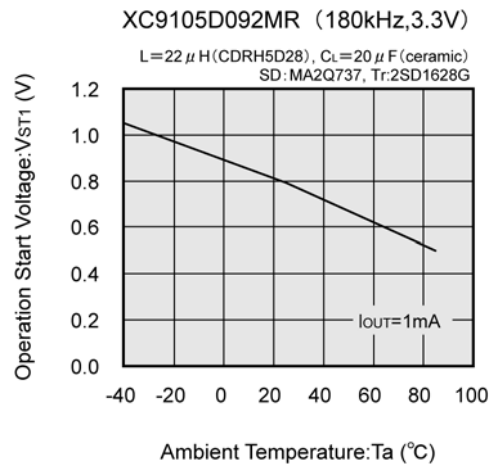
(18) Operation Hold Voltage vs. Ambient Temperature
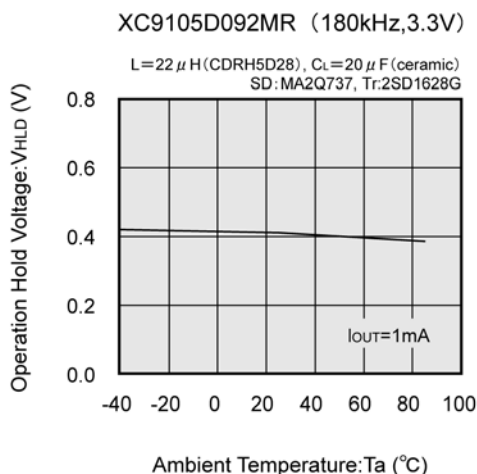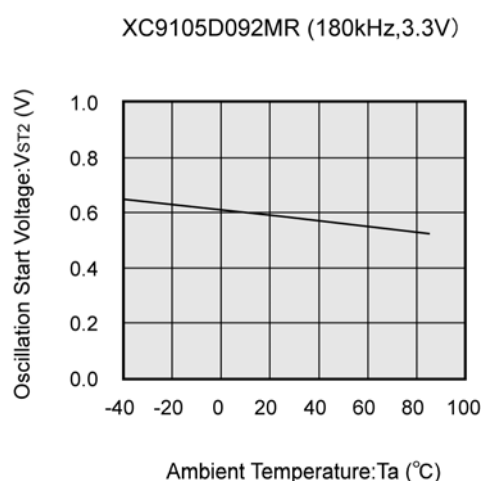
XC9105D092MR (180kHz,3.3V)

(19) Oscillation Start Voltage vs. Ambient Temperature

XC9105D092MR (180kHz,3.3V)

(20) Supply Current 1 vs. Power Supply Voltage

XC9105D093/095MR(300,500kHz)

(21) Supply Current 2 vs. Power Supply Voltage
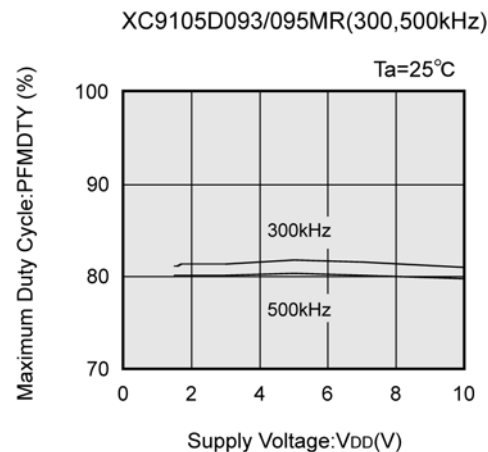
XC9105D093/095MR(300,500kHz)
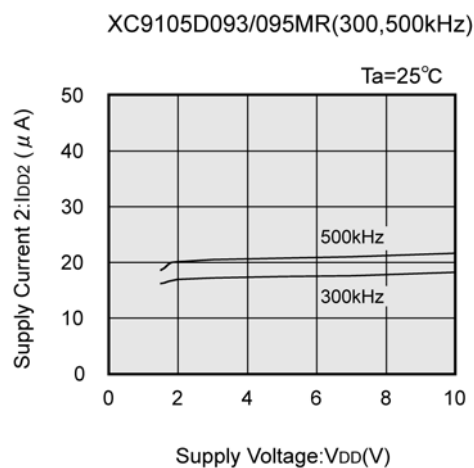
TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(22) Oscillation Frequency vs. Power Supply Voltage
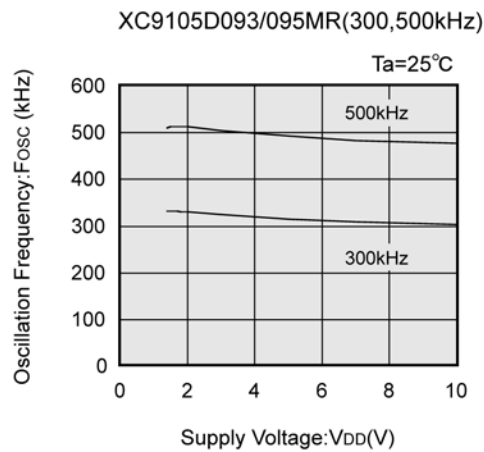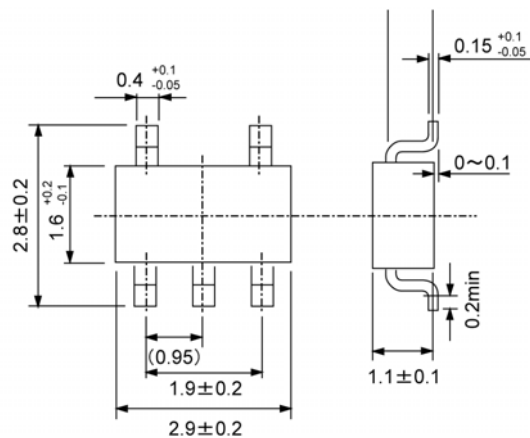
(23) Maximum Duty Cycle vs. Power Supply Voltage

XC9105D093/095MR(300,500kHz)

XC9105D093/095MR(300,500kHz)

(24) PFM Duty Ratio vs. Power Supply Voltage
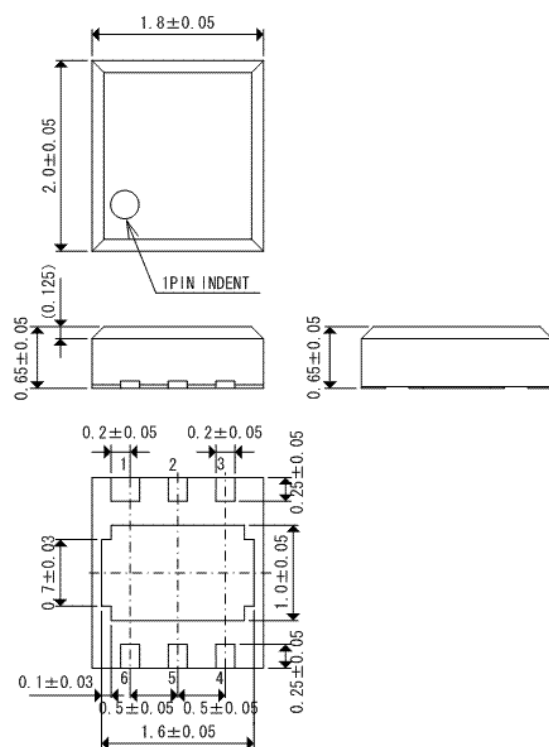

XC9105D093/095MR(300,500kHz)
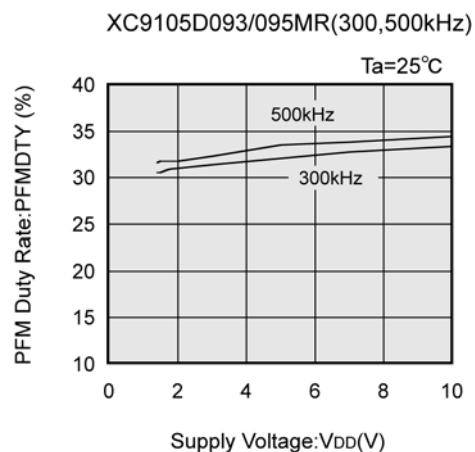
PACKAGING INFORMATION

SOT-25
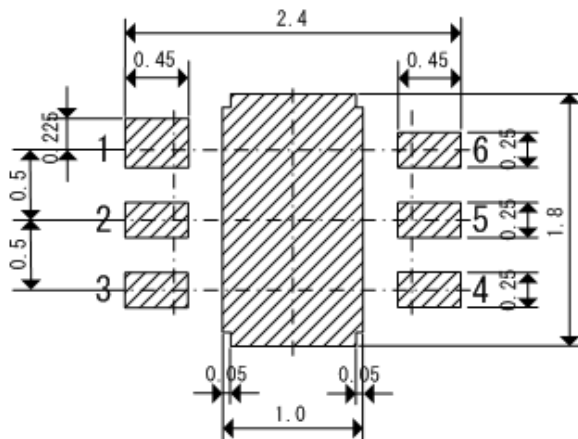
USP-6B

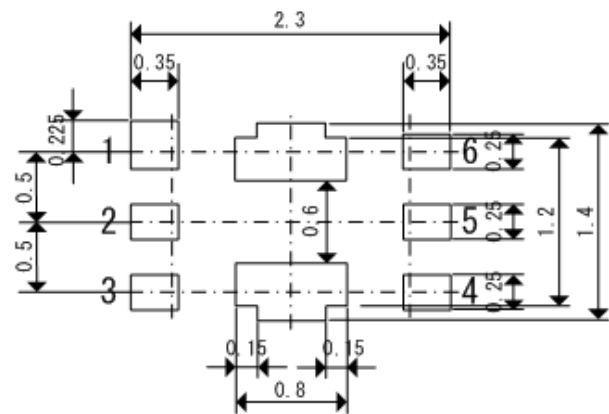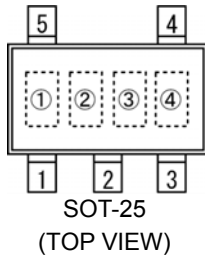## PACKAGING INFORMATION (Continued)

USP-6B Reference Pattern Layout

USP-6B Reference Metal Mask Design

MARKING RULE

SOT-25

SOT-25
(TOP VIEW)

represents product series

MARK	PRODUCT SERIES
3	XC9103x09xMx
4	XC9104x09xMx
5	XC9105x09xMx

represents current limit function

MARK	FUNCTIONS	PRODUCT SERIES
B	With current limit function	XC9103/9104/9105B09xMx
D	Without current limit function	XC9103/9104/9105D09xMx

represents oscillation frequency

MARK	OSCILLATION FREQUENCY	PRODUCT SERIES
1	100	XC9103/9104/9105x091Mx
2	180	XC9103/9104/9105x092Mx
3	300	XC9103/9104/9105x093Mx
5	500	XC9103/9104/9105x095Mx

. represents production lot number
0 to 9 and A to Z, reversed character of 0 to 9 and A to Z repeated.
(G, I, J, O, Q, W excluded)

USP-6B

USP-6B
(TOP VIEW)

. represents product series

MARK	PRODUCT SERIES
6	XC9103x09xDx
Y	XC9104x09xDx
9	XC9105x09xDx

. represents current limit function

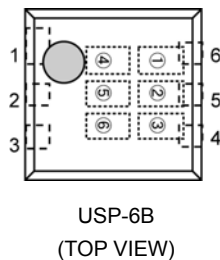| MARK	FUNCTIONS	PRODUCT SERIES
B	With current limit function	XC9103/9104/9105B09xDx
D	Without current limit function	XC9103/9104/9105D09xDx

represents FB voltage value

MARK		FB VOLTAGE	PRODUCT SERIES
0	9	09	XC9103/9104/9105x09xDx

represents oscillation frequency

MARK	OSCILLATION FREQUENCY	PRODUCT SERIES
1	100	XC9103/9104/9105x091Dx
2	180	XC9103/9104/9105x092Dx
3	300	XC9103/9104/9105x093Dx
5	500	XC9103/9104/9105x095Dx

. represents production lot number
0 to 9 and A to Z repeated. (G, I, J, O, Q, W excluded)
Note: No character inversion used.

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    (e.g. Atomic energy; aerospace; transport; combustion and associated safety equipment thereof.)

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